ELC 4351: Digital Signal Processing

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April 6, 2017

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Frequency Analysis of Signals II

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Frequency-Domain and Time-Domain Signal Properties Frequency-Domain and Time-Domain Signal Properties

Properties of the Fourier Transform for Discrete-Time Signals

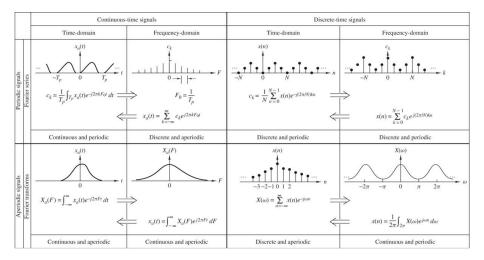
- Symmetry Properties of the Fourier Transform
- Fourier Transform Theorems and Properties

Frequency Analysis Tools

The Fourier series	for continuous-time periodic signals
The Fourier transform	for continuous-time aperiodic signals
The Fourier series	for discrete-time periodic signals
The Fourier transform	for discrete-time aperiodic signals

Continuous-time signals have aperiodic spectra Discrete-time signals have periodic spectra Periodic signals have discrete spectra Aperiodic finite energy signals have continuous spectra

The Fourier Series for Continuous-Time Periodic Signals



Periodicity with period α in one domain implies discretization with spacing $1/\alpha$ in the other domain, and *vice versa*.

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Notation

$$X(\omega) \triangleq F\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)e^{-j\omega n}$$
$$x(n) \triangleq F^{-1}\{X(\omega)\} = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega)e^{j\omega n} d\omega$$

Fourier transform pair: $x(n) \longleftrightarrow^F X(\omega)$

where, $X(\omega)$ is periodic with period 2π .

If signal is complex, it can be expressed in rectangular form

$$\begin{array}{rcl} x(n) &=& x_R(n) + j x_I(n) \\ X(\omega) &=& X_R(\omega) + j X_I(\omega) \end{array}$$

When a signal satisfies some symmetry properties in the time domain, these properties impose some symmetry conditions on its Fourier transform.

Using the rectangular form and $e^{j\omega} = \cos \omega + j \sin \omega$, we have

$$X_R(\omega) = \sum_{n=-\infty}^{\infty} [x_R(n) \cos \omega n + x_I(n) \sin \omega n]$$

$$X_I(\omega) = -\sum_{n=-\infty}^{\infty} [x_R(n) \sin \omega n - x_I(n) \cos \omega n]$$

and

$$x_R(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} [X_R(\omega) \cos \omega n - X_I(\omega) \sin \omega n] d\omega$$

$$x_I(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} [X_R(\omega) \sin \omega n + X_I(\omega) \cos \omega n] d\omega$$

Real signals. $x_R(n) = x(n)$ and $x_I(n) = 0$.

$$X_{R}(\omega) = \sum_{n=-\infty}^{\infty} x(n) \cos \omega n$$
$$X_{I}(\omega) = -\sum_{n=-\infty}^{\infty} x(n) \sin \omega n$$

It follows that

$$\begin{array}{rcl} X_R(-\omega) &=& X_R(\omega) \\ X_I(-\omega) &=& -X_I(\omega) \end{array}$$

 $\implies X^*(\omega) = X(-\omega)$. The spectrum of a real signal has *Hermitian* symmetry.

Real signals. $x_R(n) = x(n)$ and $x_I(n) = 0$.

$$egin{array}{rcl} X_R(-\omega) &=& X_R(\omega) & ({
m even}) \ X_I(-\omega) &=& -X_I(\omega) & ({
m odd}) \end{array}$$

$$|X(-\omega)| = |X(\omega)|$$
 (even)
 $\angle X(-\omega) = -\angle X(\omega)$ (odd)

Real and even signals. $x_R(n) = x(n)$, $x_I(n) = 0$ and x(-n) = x(n).

$$X_R(\omega) = x(0) + 2\sum_{n=1}^{\infty} x(n) \cos \omega n \quad (\text{even})$$
$$X_I(\omega) = 0$$

It has real-valued spectrum, which is even function of the frequency ω .

Real and odd signals. $x_R(n) = x(n)$, $x_I(n) = 0$ and x(-n) = -x(n).

$$X_R(\omega) = 0$$

$$X_I(\omega) = -2\sum_{n=1}^{\infty} x(n) \sin \omega n \pmod{2}$$

It has imaginary-valued spectrum, which is odd function of the frequency $\omega.$

Purely imaginary signals. $x_R(n) = 0$ and $jx_I(n) = x(n)$.

$$X_R(\omega) = \sum_{n=-\infty}^{\infty} x_I(n) \sin \omega n \quad (\text{odd})$$
$$X_I(\omega) = \sum_{n=-\infty}^{\infty} x_I(n) \cos \omega n \quad (\text{even})$$

Purely imaginary and odd signals. $x_R(n) = 0$, $jx_I(n) = x(n)$ and $x_I(-n) = -x_I(n)$.

$$X_R(\omega) = 2\sum_{n=1}^{\infty} x_I(n) \sin \omega n \quad (\text{odd})$$
$$X_I(\omega) = 0$$

It has real-valued spectrum, which is odd function of the frequency $\omega.$

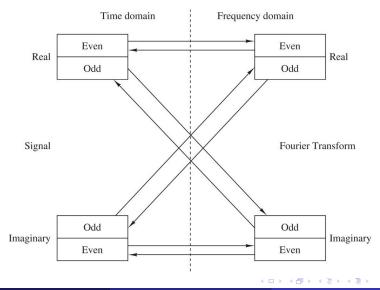
Purely imaginary and even signals. $x_R(n) = 0$, $jx_I(n) = x(n)$ and $x_I(-n) = x_I(n)$.

$$X_R(\omega) = 0$$

$$X_I(\omega) = x_I(0) + 2\sum_{n=1}^{\infty} x_I(n) \cos \omega n \quad (\text{even})$$

It has imaginary-valued spectrum, which is even function of the frequency $\omega.$

Summary of symmetry properties for the Fourier Transform



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Linearity.

If
$$x_1(n) \longleftrightarrow X_1(\omega)$$
 and $x_2(n) \longleftrightarrow X_2(\omega)$,
then
 $\alpha_1 x_1(n) + \alpha_2 x_2(n) \longleftrightarrow \alpha_1 X_1(\omega) + \alpha_2 X_2(\omega)$.

Time shifting.

If $x(n) \longleftrightarrow X(\omega)$, then $x(n-k) \longleftrightarrow e^{-j\omega k} X(\omega)$. Time reversal.

If $x(n) \longleftrightarrow X(\omega)$, then $x(-n) \longleftrightarrow X(-\omega)$. Convolution theorem.

If
$$x_1(n) \longleftrightarrow X_1(\omega)$$
 and $x_2(n) \longleftrightarrow X_2(\omega)$,
then
 $x(n) = x_1(n) \otimes x_2(n) \longleftrightarrow X(\omega) = X_1(\omega)X_2(\omega)$.

Correlation theorem.

If $x_1(n) \longleftrightarrow X_1(\omega)$ and $x_2(n) \longleftrightarrow X_2(\omega)$, then $r_{x_1x_2}(l) \longleftrightarrow S_{x_1x_2}(\omega) = X_1(\omega)X_2(-\omega).$ The Wiener-Khintchine theorem.

If x(n) is a real signal, then $r_{xx}(l) \longleftrightarrow S_{xx}(\omega)$.

Notice that neither the autocorrelation nor the energy spectral density has any phase information.

Frequency shifting.

If $x(n) \longleftrightarrow X(\omega)$, then $e^{j\omega_0 n} x(n) \longleftrightarrow X(\omega - \omega_0)$. The modulation theorem.

If $x(n) \longleftrightarrow X(\omega)$, then $x(n) \cos \omega_0 n \longleftrightarrow \frac{1}{2} [X(\omega + \omega_0) + X(\omega - \omega_0)].$ Parseval's theorem.

If
$$x_1(n) \longleftrightarrow X_1(\omega)$$
 and $x_2(n) \longleftrightarrow X_2(\omega)$, then

$$\sum_{n=-\infty}^{\infty} x_1(n) x_2^*(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(\omega) X_2^*(\omega) d\omega.$$

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Windowing theorem.

If
$$x_1(n) \longleftrightarrow X_1(\omega)$$
 and $x_2(n) \longleftrightarrow X_2(\omega)$, then

$$x_1(n)x_2(n) = rac{1}{2\pi}\int_{-\pi}^{\pi}X_1(\lambda)X_2(\omega-\lambda)d\lambda.$$

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Differentiation in the frequency domain.

If $x(n) \longleftrightarrow X(\omega)$, then

$$nx(n) \longleftrightarrow j \frac{dX(\omega)}{d\omega}.$$