# ELC 4351: Digital Signal Processing

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#### Discrete Fourier Transform

1 Fourier Transform and Discrete Fourier Transform

2 Fast Fourier Transform Algorithms

# The Fourier Transform of Discrete-Time Aperiodic Signals

### Analysis Equation

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n)e^{-j\omega n}, \qquad \omega \in [-\pi,\pi) \text{ or } \omega \in [0,2\pi)$$

#### Synthesis Equation

$$x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) e^{j\omega n} d\omega$$

 $X(\omega)$  is periodic with period  $2\pi$ .

# The Discrete Fourier Transform (DFT)

N-point DFT.

#### **Analysis Equation**

$$X(k) = \sum_{n=0}^{N-1} x(n)e^{-j2\pi \frac{k}{N}n}, \qquad k = 0, 1, 2, \dots, N-1$$

#### Synthesis Equation

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{j2\pi \frac{k}{N}n}, \qquad n = 0, 1, 2, \dots, N-1$$

 ${\it N}$  samples of the Fourier transform at  ${\it N}$  equally spaced frequencies.

$$\omega_k = \frac{2\pi k}{N}, \ k = 0, 1, 2, \dots, N - 1.$$



### **Analysis Equation**

$$X(k) = \sum_{n=0}^{N-1} x(n)W_N^{kn}, \qquad k = 0, 1, 2, \dots, N-1$$

#### Synthesis Equation

$$X(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn}, \qquad n = 0, 1, 2, \dots, N-1$$

where,  $W_N = e^{-j2\pi/N}$ 

### **Analysis Equation**

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}, \qquad k = 0, 1, 2, \dots, N-1$$

To calculate one frequency sample (each k) in the analysis equation (direct Fourier transform), we need N complex multiplications and N-1 complex additions.

For all N frequency samples, we need a total of  $N^2$  complex multiplications and N(N-1) complex additions.

### **Analysis Equation**

$$X(k) = \sum_{n=0}^{N-1} x(n)W_N^{kn}, \qquad k = 0, 1, 2, \dots, N-1$$

For all N frequency samples, we need a total of  $4N^2$  real multiplications and N(4N-2) real additions.

### **Analysis Equation**

$$X(k) = \sum_{n=0}^{N-1} x(n)W_N^{kn}, \qquad k = 0, 1, 2, \dots, N-1$$

The computation complexity of the DFT is proportional to  $N^2$ .

As a comparison, the computational complexity of the FFT is proportional to  $N \log N$ .

# Periodicity of $W_N^{kn}$

Use the periodicity of the sequence  $W_N^{kn}$  to reduce computation.

$$W_N^{kN} = e^{-j\frac{2\pi}{N}kN} = e^{-j2\pi k} = 1$$

$$W_N^{k(N-n)} = W_N^{-kn} = (W_N^{kn})^*$$
(complex conjugate symmetry)

$$W_N^{kn} = W_N^{k(n+N)} = W_N^{(k+N)n}$$
(periodicity)

Considering N an integer power of 2, i.e.  $N = 2^{\nu}$ .

$$X(k) = \sum_{n=0}^{N-1} x(n)W_N^{nk}, \quad k = 0, 1, \dots, N-1$$

$$= \sum_{n \text{ even}} x(n)W_N^{nk} + \sum_{n \text{ odd}} x(n)W_N^{nk}$$

$$= \sum_{r=0}^{N/2-1} x(2r)W_N^{2rk} + \sum_{r=0}^{N/2-1} x(2r+1)W_N^{(2r+1)k}$$

$$= \sum_{r=0}^{N/2-1} x(2r)(W_N^2)^{rk} + W_N^k \sum_{r=0}^{N/2-1} x(2r+1)(W_N^2)^{rk}$$

$$W_N^2 = e^{-2j\frac{2\pi}{N}} = e^{-j\frac{2\pi}{N/2}} = W_{N/2}$$

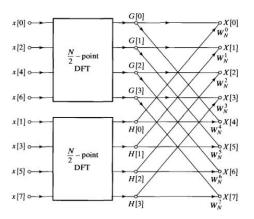
Therefore,

$$X(k) = \sum_{r=0}^{N/2-1} x(2r)(W_N^2)^{rk} + W_N^k \sum_{r=0}^{N/2-1} x(2r+1)(W_N^2)^{rk}$$

$$= \underbrace{\sum_{r=0}^{N/2-1} x(2r)W_{N/2}^{rk}}_{G(k)} + W_N^k \underbrace{\sum_{r=0}^{N/2-1} x(2r+1)W_{N/2}^{rk}}_{H(k)}$$

$$= G(k) + W_N^k H(k)$$

G(k) is an (N/2)-point DFT of even samples x(2r); H(k) is an (N/2)-point DFT of odd samples x(2r+1).

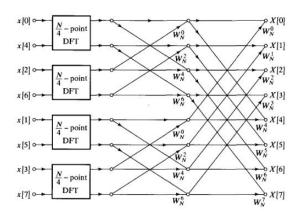


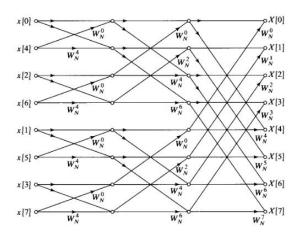
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$$G(k) = \sum_{l=0}^{N/4-1} g(2l) W_{N/4}^{lk} + W_{N/2}^{k} \sum_{l=0}^{N/4-1} g(2l+1) W_{N/4}^{lk}$$

$$H(k) = \sum_{l=0}^{N/4-1} h(2l) W_{N/4}^{lk} + W_{N/2}^{k} \sum_{l=0}^{N/4-1} h(2l+1) W_{N/4}^{lk}$$

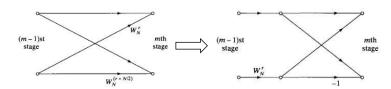
$$(N/4)-\text{point DFT}$$

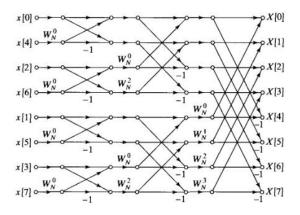




# Symmetry of $W_N$

$$egin{array}{lcl} W_{N}^{N/2} & = & e^{-jrac{2\pi}{N}rac{N}{2}} = e^{-j\pi} = -1 \ W_{N}^{r+N/2} & = & W_{N}^{N/2}W_{N}^{r} = -W_{N}^{r} \end{array}$$





# Bit-reverse Reordering

x(n)'s index n	binary
0	000
4	100
2	010
6	110
1	001
5	101
3	011
7	111

# Decimation-in-Frequency Fast Fourier Transform (FFT)

